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SINGLE SPACE SEGMENT

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HRPT / LRPT DIRECT BROADCAST SERVICES SPECIFICATION



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DOCUMENT CHANGE RECORD

Issue / Revision	Date:	DCN No.	Changed Pages / Paragraphs
Issue 1	13/06/95		
Issue 2	15/05/96		All pages
Issue 3	04/02/97		All pages
Issue 4	14/11/97		page 7: GOME packetised data rate increased to 400 kbit/s max; Satellite housekeeping packet and GRAS pos. & timing packet size adjusted to reflect the constraint of the size being multiple of 32 octets; DCS-2 packet data rate modified. page 14: data rate budget updated page 30: The HRPT EIRP has been decreased by 1 dB over the whole coverage. page 36/37: The HRPT link budget has been modified to consider the variation in the HRPT EIRP.







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Issue 5 Daft	01/02/98	MO-DS-ESA-SY0048 EPS/SYS/SPE/95413	page 1: reference to document AD01 removed; cosmetic. page 2: reference to AD01 and RD01 removed. page 3: compression scheme referred to Annex 2 page 4: packet structure format modified; Application process identifier set always to 1; Sequence flag set always to 11 page 5: Packet secondary header description modified; Ancillary data field description page 6: packet data rate table updated page 7: reference to RD01 removed page 11: insert zone is always present; cosmetic page 13: data rate budget table updated page 14/15/16: LRPT physical layer description modified page 20: level required to demodulate LRPT data stream modified; cosmetic page 31/33: link budget updated; cosmetic page 31/33: link budget updated; cosmetic page 35: depth of compression and selection of transmitted channels modified page 38: packet sequence modified page 39: packet structure modified page 41: packet structure modified page 43: cosmetic
Issue 5	1/10/98		page 5: update to UTC and SBT field description; page 15: identification of sync marker in place of TBD. page 36-40: detailed changes for compatibility with 'Reference Compressor/Decompressor'



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1. INTRODUCTION

This document defines the specification of the HRPT and LRPT services provided by the METOP satellite.

The HRPT and LRPT direct broadcast services are characterised by the implementation of the CCSDS Recommendations (AD03 and AD04) and the encryption of selectable virtual channels. The on-board encryption mechanism, based on the DES algorithm, is described in [AD02]. The encryption is addressed in this document, only to specify the content of the Insert zone in the VCDU data structure in the Data Link Layer.

This document covers the implementation of the data communication model on METOP spacecrafts.

The structure of this document follows is as follows:

Chapter 1 - This chapter

Chapter 2 - Lists the applicable and reference documents

Chapter 3 - Details the Application Layer implementation specific to the EPS /

METOP mission.

Chapter 4 - Deals with the Network Layer implementation details.

Chapter 5 - Deals with the Data Link Layer implementation details.

Chapter 6 - Describes the LRPT Physical Layer and the Ground Station main

requirements.

Chapter 7 - Describes the HRPT Physical Layer and the Ground Station main

requirements.

Annex 1 - Provides an overview of the link budget for HRPT and LRPT.

Annex 2 - Describe the algorithm used to compress the AVHRR High Rate and the

packet format of the resulting AVHRR Low rate application.

Annex 3 - Lists the acronyms used in this document.



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2. DOCUMENTATION

2.1 Applicable Documents

AD01 Deleted

AD02 EPS/SYS/SPE/95424 EUMETSAT Polar System / METOP Programme:

MO/DS/ESA/SY/0049 Encryption System Specification

AD03 CCSDS 701.0-B-2 Advanced Orbiting Systems, Networks and Data, Blue Book,

Issue 2.

AD04 CCSDS 101.0-B-3 Telemetry Channel Coding, Blue Book, Issue 3.

2.2 Reference Documents

RD01 Deleted



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3. APPLICATION LAYER

The Application layer defines the information exchange between the METOP payload and the user specific applications.

The NOAA procured instruments (AVHRR, AMSU-A1, AMSU-A2, HIRS, SEM, DCS) will generate raw data which will be time tagged and formatted by the spacecraft NOAA Interface Unit (NIU). This NIU will also provide data compression of the AVHRR data for transmission on the LRPT link.

The IASI, MHS, GRAS, GOME and ASCAT instruments provide data in the form of CCSDS source packets.

The satellite provides housekeeping data, GRAS positioning and timing data and administrative messages in the form of CCSDS source packets.

The Application Process Identifiers relevant to each instrument are defined in 3.3.

3.1 Application Data

3.1.1 Application Data provided with LRPT

The application data provided by the Low Resolution Picture Transmission link are as follows:

- Compressed resolution imagery on selected channels of the AVHRR instrument (the data compression scheme is defined in Annex 2).
- Infrared and microwave sounding data from the Meteorological Payload: AMSU-A1, AMSU-A2, MHS, HIRS.
- SEM data.
- Spacecraft Housekeeping data.
- GRAS positioning and timing data.
- Administrative messages.

3.1.2 Application Data provided with HRPT

The application data provided by the High Resolution Picture Transmission link are as follows:

- Full resolution AVHRR imagery.
- Infrared and microwave sounding data from the Meteorological Payload: AMSU-A1, AMSU-A2, MHS, HIRS and IASI.
- SEM data.
- DCS data.
- Data provided by ASCAT, GOME.
- Spacecraft Housekeeping data.
- GRAS positioning and timing data.



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- GRAS sounding data.
- Administrative messages.

3.2 Source Packet structure

The source packet, in addition to the source data carries information needed for the acquisition, storage, distribution and exploitation of the source data by the end user.

The source packet structure is as follows:

	Packet Primary Header (48 bits)								User data	
	Packet identifier 2 octets				ience control	Packet length 2 octets	8 octets	8 octets variable		
Version No	Type	Secondary Header Flag 1 bit	APID	Sequence flag	Packet sequence Count		Time stamp	Ancillary data	Application data	PEC
3 bits "000"	1 bit "0"		11 bits	2 bits	14 bits	16 bits	64 bits	var	var	16 bits

The utilisation of the fields within the primary header is as follows:

Packet identifier

Version number
 Type
 000 (CCSDS packet Version number 1)
 (This bit is not used within AOS)

- Secondary Header Flag This bit shall be always set to 1 to indicate the presence of a

secondary header.

- Application Process Identifier This field defines the data route between two users application

endpoints: the APIDs are listed in 3.3.

Packet sequence control

- Sequence Flag This flag is set to 11 indicating that the packet contains

unsegmented User data.

The maximum length of the packet is 65542 octets.

- Packet name/sequence count This field is a modulo 16384 counter, which numbers the

packets.

- Packet length This field contains a sequential binary count "C" that expresses

the length of the Secondary Header and the User Data. The

value of "C" is the length (in octets) minus 1.



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The Packet secondary header contains the time stamp. The time stamp is associated to a known time preceding the event measured. The time stamp is compliant with CCSDS 301-B2 "level 1" Time Code.

The time stamp consists of

- 2 octets indicating the number of days with reference to 1/1/2000;
- 4 octets indicating the millisecond of the day;
- 2 octets indicating the microsecond of the millisecond.

The time stamp will be synchronised to UTC with an accuracy of 4 milliseconds.

The User Data field contains the following fields:

Ancillary data field: this field contains in the first six octets a secondary time stamp and, optionally, other information required for the processing of the application data - i.e. instrument mode, instrument telemetry and calibration data, redundancy -. Its size - an even number of octets - depends on the instrument requirement.

The secondary time stamp consists of:

- one octet filled with "0";
- 3 octets of coarse time (second);
- 2 octets of fine time (2⁻¹⁶ second). Within this field, a number of the LSB's (depending on the source) up to a maximum of 8 may be set to "0".

The reference time of the secondary time stamp (epoch) is known and defined by the Satellite Control Centre. A counter rollover happens every 2^{24} seconds (half a year). This is equivalent to a modification of the epoch. The secondary time stamp is synchronised with the time stamp in the Packet secondary header.

- Application data: this field contains information provided by the source; its length shall be an even number of octets.
- Packet Error Control: this field is optional: if required by the user it shall contain one of the following checksum:
 - a) a Cyclic Redundancy Checksum (CRC) computed over all other octets that constitute the packet. The polynomial generator shall be: $G(x) = x^{16} + x^{12} + x^5 + 1$.

Both encoder and decoder shall be initialised with all ones state for each packet.

b) a vertical parity checksum calculated by performing an exclusive-OR on all the other octets pairs that constitute the packet



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3.3 Application data overview: APID and Source Packets size - data rates

The following table provides an overview of the METOP application data: It shows the reserved APIDs, the packetised data rate, the size of the whole packet, the Ancillary data and the Application data length and eventually the packet rate for each application.

Application	APID	packetised data rate (kbps) average/ (peak)	Packet size (octet)	Ancillary data length (octets)	Application data length (octet)	Rate (packet /s)	
AVHRR/3 high resolution	AVHRR/3 high 103, 104 622.272		12964	6	12944	6	
AVHRR/3 low resolution	6470	40	var	12	var	3	
MHS	34	3.918	1306	6	1286	3/8	
DCS-2	35	7.5 max		1	lata length (octets) data length (octet) (packet /s) 6 12944 6 12 var 3 6 1286 3/8 tbd 6 640 1/32 6 2304 1/6.4 6 2080 1/8 6 1120 1/8 detailed in the ICD detailed in the ICD tbd tbd tbd detailed in the ICD detailed in the ICD 6 524 1 6 tbc 172 or 12 (tbc) 1/16 (tbc)		
SEM (option)	37	0.165	660			1/32	
HIRS/3	38	2.905	2324	6	2304	1/6.4	
AMSU-A1	39	2.100	2100	6	2080	1/8	
AMSU-A2	40	1.140	1140	6	1120	1/8	
IASI	128191	1500/ (2200)	detailed in the ICD				
ASCAT	192255	43 (60)		detailed	in the ICD		
GOME	384447	400 max	tbd	tbd	tbd	tbd	
GRAS meteorological data	448511	20/(60)		detailed	in the ICD		
Satellite housekeeping packet	1	4.352	544	6	524	1	
GRAS positioning 2, 3 and timing data		<0.100	192 or 32 (tbc)	6 tbc		1/16	
Administrative messages	6	2.008	8032 tbc	8	8012 tbc	1/32	

3.4 Detailed Instrument source packet description

Although the instruments data follow the CCSDS recommendation together with the specific requirements listed in chapter 3.3, the packet data field structure is specific to the instrument itself.



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4. NETWORK LAYER

The Network layer is represented by the path layer in the CCSDS standard. In this case, the only function of the path layer shall be to generate the VCDU-Id and to forward CP-PDUs to the multiplexing service.

The VCDU-ID is a data structure of 14 bits length, consisting of a spacecraft identifier (8 bits) and a virtual channel identifier (6 bits).

Spacecraft identifiers shall be assigned as follows:

Spacecraft	Identifier			
METOP1	00001011			
METOP2	00001100			
METOP3	00001101			
METOP SIMULATOR	00001110			

The virtual channel identifiers are given by the subsequent table:

LRPT:

Instruments	Virtual Channel Identifier
Spacecraft Housekeeping	34
MHS	12
AMSU A1/2, SEM, HIRS	3
AVHRR low rate	5
GRAS positioning and timing data	34
Admin message	34
Fill VC	63



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HRPT:

Instruments	Virtual Channel Identifier
Spacecraft Housekeeping	34
MHS	12
AMSU A1/2, SEM, HIRS	3
DCS	27
AVHRR High rate	9
IASI	10
ASCAT	15
GOME	24
GRAS positioning and timing data	34
GRAS sounding data	29
Adm. messages	34
fill VC	63



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5. DATA LINK LAYER

The Data Link Layer is organised into two sublayers: a Virtual Channel Link Control sublayer (VCLC) and a Virtual Channel Access sublayer (VCA). The VCLC sublayer receives CCSDS packets from the Network layer, while the VCA sublayer forwards the physical channel access protocol data unit (PCA_PDU) to the physical layer.

The virtual channel procedures are functions required to generate virtual channel data units (VCDUs) from VCA_SDUs and vice versa. One of the channel access procedures is to handle Reed-Solomon check symbols. A VCDU with attached check symbols is called coded virtual channel data unit (CVCDU). The PCA_PDU consists of a succession of CVCDU prefixed by a Synchronisation Marker.

The structure of one CVCDU is shown in the following figure:

		VCDU Primary Header (6 octets)					VCDU insert zone	VCDU Data Unit Zone			CVCDU													
	Version n°	VCD	U Id	VCDU Signalling Field counter						Signalling Field				Signalling Field		Signalling Field		Signalling Field				M_PDU header M_PDU packet zone		Check symbols
I					1	1 octet		2 octets																
		S/C id 8 bits	Type 6 bits	3 octets	Replay flag	spare	2 octets	M_PDU header	M_PDU first header pointer	882 octets	128 octets													
Į	"01"				"0"	"0000000"		spare																

The elements of the CVCDU are as follows:

VCDU primary header	contains a six octets header structure
VCDU insert zone	contains one IN_SDU having a length of 2 octets
VCDU data unit zone	contains one VCA_SDU in case of a valid VCDU or all zeros in case of a fill VCDU, the size of this field is 884 octets
Reed-Solomon check symbols	contain Reed-Solomon code (255,223) encoded check symbols, calculated over the VCDU primary header and the VCDU data unit zone.



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5.1 VCDU primary header

The VCDU primary header consists of the following elements:

version number	set to"0 1" specifying version-2 CCSDS structure
VCDU-ID	virtual channel data unit identifier as specified in Chapter 4, consisting of spacecraft identifier and virtual channel identifier.
VCDU counter	sequential count (modulo 16777216) of VCDUs on each virtual channel
signalling field	set to 0 specifying real-time VCDUs

5.2 VCDU Insert Zone

The insert zone is always present and used for encryption control (AD02).

The structure of the IN_SDU used with LRPT or HRPT is as follows:

The insert service data unit (IN_SDU) is used for data encryption: this field is composed of:

- Encryption flag (1 octet): set to $00_{\rm HEX}$ when encryption is off; set to FF $_{\rm HEX}$ when encryption is on.
- Key number (1 octet): this octet indicates which message key is used to encrypt the VC. It is set to 00_{HEX} when encryption is off.

In case of failure of the encryption mechanism, the system shall not prevent data transmission to the ground; data shall be transmitted in the clear.

5.3 VCDU Data Unit Zone

The CVCDU data unit zone contains the multiplexing protocol data unit; this field consists of:

- M_PDU Header Spare bits (5 bits) : all set to "0"
- M_PDU Header First Pointer (11 bits): it contains a binary count P, which, when incremented by one, points directly to the number of the octet that contains the first octet of the first CCSDS packet header. If the VCDU data zone does not contain any packet header at all, the bits shall be set to "1".
- M_PDU Packet Zone (882 octets): it contains part, parts or complete CCSDS packets.



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5.4 Fill VCDU

In the event that there are no valid M_PDU available for transmission, a fill VCDU will be generated. The content of the VCDU Data Unit Zone will be all "0".

5.5 Reed Solomon Check Symbol Field

The Reed Solomon Check Symbol Field contains the check symbols which allow error correction. They are generated according to AD04 with an interleaving depth of I=4.

5.6 Randomisation

Each commutated sequence of CVCDUs is converted into a sequence of channel access data units (CADUs). For this purpose each CVCDU is randomized first and preceded by a synchronization marker then.

Randomization is performed by multiplying all 8160 bits of the CVCDU with a pseudo noise pattern. The pseudo noise sequence is generated by means of the following polynomial:

$$h(x) = x^8 + x^7 + x^5 + x^3 + 1$$

This sequence repeats after 255 bits with the sequence generator being reinitialized to an all-ones state. The resulting PN pattern begins with (hexadecimal) FF480EC09A.

5.7 Synchronisation marker

The synchronization marker is defined to be (hexadecimal)

1ACFFC1D

which describes a 32 bits pattern to precede each CVCDU.

Each CADU has a length of 8192 bits.

5.8 CVCDUs commutation algorithm

Commutation of CVCDUs shall not be performed on basis of a priority system.



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5.9 Data rate budget

The data rate budget for the LRPT and HRPT channels is given for information here below:

Application	HRPT [kb/s]	LRPT [kb/s]
IASI	1500.000	0.000
AVHRR	622.272	40.000
HIRS	2.905	2.905
AMSU A1	2.100	2.100
AMSU A2	1.140	1.140
MHS	3.918	3.918
SEM	0.165	0.165
DCS	7.500	0.000
ASCAT	60.000^{1}	0.000
$GRAS^2$	60.000 ¹ (tbc)	0.000
GOME	400.000 (tbc)	0.000
Satellite housekeeping packet	4.352	4.352
GRAS time and real time position	0.100	0.100
Admin messages	2.008	2.005
subtotal	2,666.460	56.685
Capacity reserved	348.183	5.331
sub-total	3,014.643	62.016
RS-coding (+16.1%)	485.357	9.984
total	3500.000	72.000

- 1) Peak data rate
- 2) Occultation data



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6. LRPT PHYSICAL LAYER

The LRPT physical layer shall perform the following operations (see for reference the modulator block diagram in figure 6.1):

- 1) Convolutional encoding
- 2) Interleaving of the convolutionally coded signal
- 3) Insertion of a UW for interleaving synchronisation and delimitation
- 4) Serial to parallel conversion
- 5) Modulation according to the QPSK format
- 6) Amplification of the modulated signal
- 7) Transmission from the LRPT antenna

6.1 Convolutional encoding

The input data stream shall be convolutional encoded.

The characteristics of the encoder are the following

Code rate: ½
Constraint length: 7 bits

Connection vectors: G1 = 1111001 / G2 = 1011011

Symbol inversion: No Puncturing: No

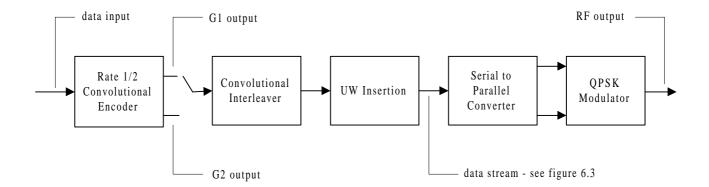


Figure 6.1 - Modulator block diagram



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6.2 Interleaving

In this section coded data units will be called bits.

The bits delivered by the convolutional encoder are shifted sequentially into a bank of registers. With each new coded bit, the commutator switches to a new register, and the new bit is shifted in while the old coded bit in that register is shifted out to the following stage (see figure 6.2).

The output G1 of the encoder shall feed the odd branches, whereas the output G2 shall feed the even branches.

The number of the interleaver branches (B) shall be 36.

The number of the elementary delay (M) in each branch shall be 2048 bits.

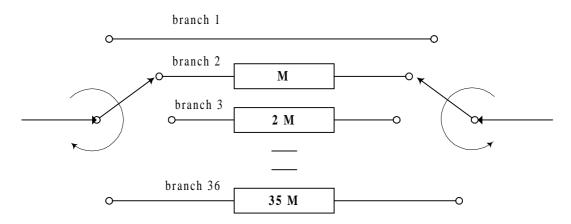


Figure 6.2 - Interleaver block diagram



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6.3 Synchronisation marker insertion

A synchronisation marker shall be inserted every 72 bits of the data stream delivered by each interleaving process. The synchronisation marker is 8 bit long and is defined as 00(i,q), 10(i,q), 01(i,q), 11(i,q).

A synchronisation marker is inserted at the output of the convolutional interleaver after the bit supplied by the last (36^{th}) branch every two frames. The frame is structured as shown in figure 6.3.

The bit rate after the synchronisation marker insertion is 160 kbit/s.

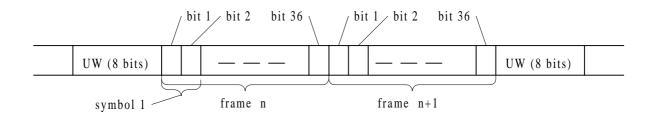


Figure 6.3 - Frame structure

6.4 Serial to parallel conversion

The grouping of pairs of bits for the QPSK modulator shall be obtained from the 1^{st} and 2^{nd} , the 3^{rd} and 4^{th} etc. branches of the convolutional interleaver output.

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6.5 QPSK Modulation

6.5.1 Modulation mapping

The mapping onto the QPSK constellation shall be according the Gray encoding (see figure 6.4).

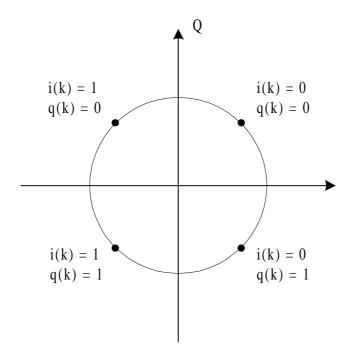


Figure 6.4 - QPSK constellation diagram

6.5.2 Modulation waveform

The data stream shall be modulated according to the QPSK format. The QPSK format can be expressed in complex notation as:

$$\tilde{s}_{T}(t) = \sum_{i=-\infty}^{\infty} \sqrt{2P_{T}} \left[\cos(\phi_{i}) g_{T}(t-iT_{s}) + j \sin(\phi_{i}) g_{T}(t-iT_{s}) \right]$$



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The baseband square-root raised-cosine filter impulse response is given by:

$$g_{T}(t) = \frac{1}{\pi} \left\{ \left[\frac{16\sqrt{T_{s}}\alpha^{2}t}{(4\alpha t + T_{s})(4\alpha t - T_{s})} - \frac{\sqrt{T_{s}}}{t} \right] \sin\left(\frac{(\alpha - 1)\pi t}{T_{s}}\right) - \frac{4\alpha\sqrt{T_{s}^{3}}\cos\left(\frac{(\alpha + 1)\pi t}{T_{s}}\right)}{(4\alpha t + T_{s})(4\alpha t - T_{s})} \right\}$$

where Ts is the symbol duration, ϕ i is the information bearing phase (ϕ i belongs to the 4-ary alphabet $\{\pm\pi/4; \pm 3\pi/4\}$), α is the roll-off factor, P_T is the transmitter power.

6.5.3 RF parameters

The QPSK ideal modulation is approximated by a two NRZ filtered data streams modulating the carrier on two orthogonal axis.

The two synchronous, NRZ rectangular pulses streams shall be filtered in baseband by a network approximating the square-root raised cosine Nyquist filter, defined as follows:

$$|H(j\omega)| = \begin{cases} \frac{\omega T_s/2}{\sin(\omega T_s/2)} & 0 \le \omega \le \frac{\pi}{T_s} (1-\alpha) \\ \frac{\omega T_s/2}{\sin(\omega T_s/2)} \cos \left[\frac{T_s}{4\alpha} \left(\omega - \frac{\pi(1-\alpha)}{T_s} \right) \right] & \frac{\pi}{T_s} (1-\alpha) \le \omega \le \frac{\pi}{T_s} (1+\alpha) \\ 0 & \omega \ge \frac{\pi}{T_s} (1+\alpha) \end{cases}$$



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The Segment delay caused by the filtering network shall be within the mask shown in figure 6.5 / Tables 6.1 and 6.2.

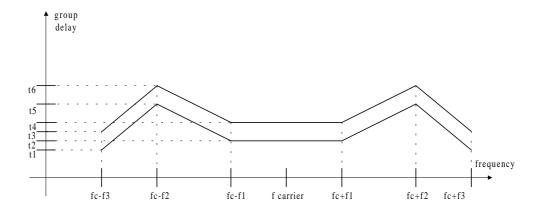


Figure 6.5 - Segment delay of the transmitting channel

Segment delay (nsec)	
t1	tbd
t2	tbd
t3	tbd
t4	tbd
t5	tbd
t6	tbd

Frequency		
f1	tbd	
f2	tbd	
f3	tbd	
f4	tbd	

Table 6.1 Table 6.2

This Segment delay distortion shall be compensated for at the receiving end to obtain the minimum channel degradation.



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The roll-off factor shall be 0.6.

The LRPT physical layer shall generate a signal which requires an Eb/No of 4.5 dB to be coherently demodulated with a BER of 10-3 at the output of the Viterbi decoder.

The loss introduced by the demodulator (used for the test) shall be considered part of the requirement.

The loss associated to the synchronisation insertion - which is $10 \cdot \log(80/72)$ dB; 0.5 dB - shall not be taken into account in this figure.

This requirement shall be met at the maximum allowed compression point of the amplifier and for a random data pattern.

The nominal carrier frequency shall be either: 137.1 MHz or 137.9125 MHz

The bandwidth of the signal (99% of total power) shall not exceed 150 kHz.

The carrier frequency deviation from the nominal or back-up frequency, including initial accuracy and drift due to aging and temperature, shall not exceed $\pm 15 \cdot 10$ -6.

6.6 Amplification

The QPSK modulated signal shall be amplified in order to achieve the EIRP defined in para 6.7.

The working point of the amplifier shall be selected in order to meet the requirements defined in para 6.5.3.



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6.7 Transmission

The LRPT subsystem shall at least radiate at the S/C interface the EIRP defined in table 6.3.

Angle w.r.t. Nadir	EIRP (dBW)
0	3.21
5	3.24
10	3.36
15	3.55
20	3.82
25	4.19
30	4.65
35	5.23
40	5.95
45	6.85
50	8.00
55	8.00
60	8.00
62	8.00

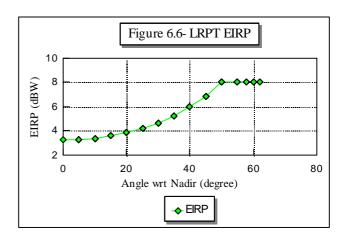


Table 6.3

The polarisation shall be RHCP.

The axial ratio shall be less than 4.5 dB.



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6.8 LRPT ground stations

The LRPT link shall serve ground station equipments of two different types:

- Steerable antennas with a minimum elevation angle of 5°
- Omnidirectional antennas with a minimum elevation angle of 13°

The assumed ground station G/T and corresponding minimum elevation angles (used for link margin calculation provided in Annex 1) are as follows:

Antenna type	System G/T	min. S/C Elevation
steerable YAGI antenna	-22.4 dB/K	5°
omni-directional antenna	-30.4 dB/K	13°

Axial ratio: less than 3.5 dB.

Pointing losses: less than 0.2 dB (YAGI antenna).

The assumptions taken on ground stations modulation and receiver degradation for link budget calculation, are given in Annex 1.



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7. HRPT PHYSICAL LAYER

The HRPT physical layer shall perform the following operations:

- 1) Convolutional encoding
- 2) Modulation according to the QPSK format
- 3) Amplification of the modulated signal
- 4) Transmission from the HRPT antenna

7.1 Convolutional encoding

The input data stream shall be convolutional encoded.

The characteristics of the Viterbi encoder shall be the following

Code rate: 3/4

Constraint length: 7 bits

Connection vectors: G1= 1111001 / G2=1011011

Phase relationship: G1 is associated with the first symbol

Symbol inversion: No

Puncturing: Yes

Puncturing scheme:

- The 3/4 rate code is realised by puncturing the output of a $\frac{1}{2}$ rate Viterbi coder (see figure 7.1).
- The output streams from the 3/4 rate Viterbi encoder labelled i(k) and q(k) consist of the output streams of the ½ rate Viterbi coder labelled l(k) and m(k) and associated with the G1 and G2 vectors with the exception of two out of six bits, which are deleted in a repeating pattern.



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- The bits to be deleted are shown struck out:

- Therefore the two streams i(k) and q(k) are composed by the following bits:

$$i(k) = ..., l(k), l(k+2), l(k+3), l(k+5), l(k+6),$$

$$q(k) = ..., m(k), m(k+1), m(k+3), m(k+4), m(k+6),$$

The output of the Viterbi encoder has a rate of 4666.667 kbit/s and is provided to the modulation section.

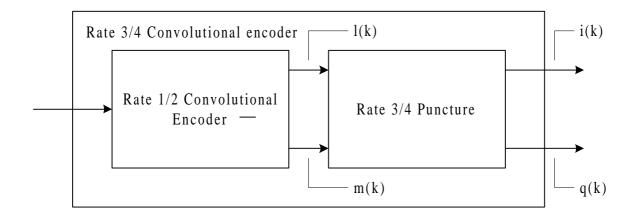


Figure 7.1 - Rate 3/4 convolutional encoder



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7.2 QPSK Modulation

7.2.1 Modulation mapping

The mapping onto the QPSK constellation shall be according the Gray encoding (see figure 7.1).

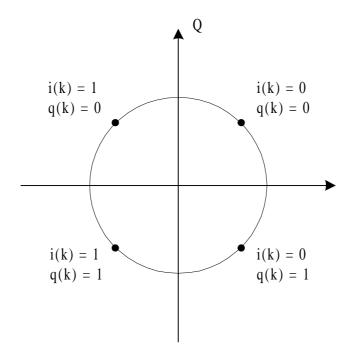


Figure 7.2 - QPSK constellation diagram

7.2.2 Modulation Waveform

The data stream shall be modulated according to the QPSK format. The QPSK format can be expressed in complex notation as:

$$\tilde{s}_T(t) = \sum_{i=-\infty}^{\infty} \sqrt{2P_T} \left[\cos(\phi_i) g_T(t - iT_s) + j \sin(\phi_i) g_T(t - iT_s) \right]$$



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The baseband square-root raised-cosine filter impulse response is given by:

$$g_{T}(t) = \frac{1}{\pi} \left\{ \left[\frac{16\sqrt{T_{s}}\alpha^{2}t}{(4\alpha t + T_{s})(4\alpha t - T_{s})} - \frac{\sqrt{T_{s}}}{t} \right] \sin\left(\frac{(\alpha - 1)\pi t}{T_{s}}\right) - \frac{4\alpha\sqrt{T_{s}^{3}}\cos\left(\frac{(\alpha + 1)\pi t}{T_{s}}\right)}{(4\alpha t + T_{s})(4\alpha t - T_{s})} \right\}$$

where Ts is the symbol duration, ϕ i is the information bearing phase (ϕ i belongs to the 4-ary alphabet $\{\pm \pi/4; \pm 3\pi/4\}$), α is the roll-off factor, P_T is the transmitter power.

7.2.3 RF parameters

The QPSK ideal modulation is approximated by a two NRZ filtered data streams modulating the carrier on two orthogonal axis.

The two synchronous, NRZ rectangular pulses streams shall be filtered in baseband by a network approximating the square-root raised cosine Nyquist filter, defined as follows:

$$|H(j\omega)| = \begin{cases} \frac{\omega T_s/2}{\sin(\omega T_s/2)} & 0 \le \omega \le \frac{\pi}{T_s} (1-\alpha) \\ \frac{\omega T_s/2}{\sin(\omega T_s/2)} \cos \left[\frac{T_s}{4\alpha} \left(\omega - \frac{\pi(1-\alpha)}{T_s} \right) \right] & \frac{\pi}{T_s} (1-\alpha) \le \omega \le \frac{\pi}{T_s} (1+\alpha) \\ 0 & \omega \ge \frac{\pi}{T_s} (1+\alpha) \end{cases}$$



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The Segment delay caused by the filtering network shall be within the mask shown in figure 7.3 / Tables 7.1 and 7.2.

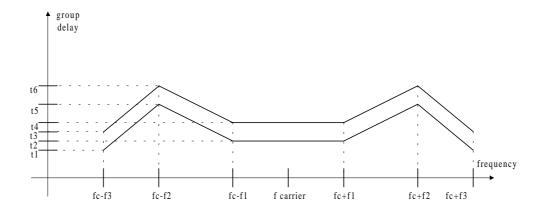


Figure 7.4 - Transmission channel Segment delay mask

Segment delay (nsec)		
t1	tbd	
t2	tbd	
t3	tbd	
t4	tbd	
t5	tbd	
t6	tbd	

Frequency		
f1	tbd	
f2	tbd	
f3	tbd	

Table 7.1 Table 7.2

This Segment delay distortion shall be compensated for at the receiving end to obtain the minimum channel degradation.

The roll-off factor shall be 0.6.

The HRPT physical layer shall generate a signal which requires an Eb/No of 5.0 dB to be coherently demodulated with a BER of 10-3 at the output of the Viterbi decoder. The loss introduced by the demodulator (used for the test) shall be considered part of the requirement.



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This requirement shall be met at the maximum allowed compression point of the amplifier and for a random data pattern.

The Nominal carrier frequency shall be:

either 1701.300 MHz, being the nominal bandwidth 1698.75 - 1703.25 MHz

or 1707.000 MHz; being the nominal bandwidth 1704.75 - 1709.25 MHz.

The nominal bandwidth shall contain 99% of the total signal power.

The carrier frequency deviation from the nominal or back-up frequency, including initial accuracy and drift due to aging and temperature, shall not exceed $\pm 25 \cdot 10$ -6.

7.3 Amplification

The QPSK modulated signal shall be amplified in order to achieve the EIRP defined in para 3.4. The working point of the amplifier shall be selected in order to meet the requirements defined in para 7.2.3.



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7.4 Transmission

The HRPT subsystem shall at least radiate at the S/C interface the EIRP defined in table 7.3.

Angle w.r.t. nadir	EIRP (dBW)
0	1.46
5	1.49
10	1.61
15	1.8
20	2.07
25	2.44
30	2.9
35	3.48
40	4.2
45	5.1
50	6.25
55	7.8
58	9.1
60	9.1
62	9.1

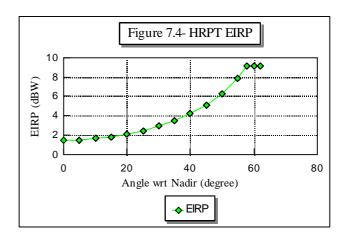


Table 7.3

The gain of the antenna shall be such that the following PFD limitations are met:

- -154 dBW/(m^2 ·4 kHz) for elevation angle (δ) lower than 5°
- -154 + 0.5 (δ -5) dBW/(m²·4 kHz) for elevation angle (δ) between 5° and 25°
- -144 for elevation angle (δ) greater than 25°
- -133 dBW/(m²·1.5 MHz) at any elevation angle (from 1670 to 1700 MHz).

The following assumptions shall be used in the PFD calculation:

- Peak of TX spectrum density including possible residual carrier.
- Typical values for TX power and antenna gain over elevation.
- Average values for antenna gain along azimuth.

The polarisation shall be RHCP. The axial ratio shall be less than 4.5 dB.



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7.5 HRPT ground stations

The HRPT link shall serve ground stations, which may be located anywhere in the world, and which will have the following characteristics:

Ground station axial ratio less than 1 dB.

The assumptions taken on ground stations modulation and receiver degradation for link budget calculation, are given in Annex 1.



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Annex 1. TYPICAL WORST CASE LRPT/HRPT LINK BUDGETS BASED ON SPECIFICATION VALUES

Link ID	METOP LRPT - High gain antenna receiver								
Ground station	LRPT (G/T=-22.4)								
	I								
Frequency		(MHz)	137.90	137.90	137.90	137.90	137.90	137.90	137.90
S/C altitude		(km)	850.00	850.00	850.00	850.00	850.00	850.00	850.00
Slant range		(km)	2,889	2,468	1,858	1,473	1,227	963	850
S/C view angle		(degree)	61.51	60.33	56.00	49.82	42.52	26.18	0.00
Data rate		(kbit/s)	72.00	72.00	72.00	72.00	72.00	72.00	72.00
G/S view angle		(degree)	5.00	10.00	20.00	30.00	40.00	60.00	90.00
S/C EIRP	note 1	(dBW)	8.00	8.00	8.00	7.96	6.40	4.28	3.21
S/C antenna axial ratio		(dB)	4.50	4.50	4.50	4.50	4.50	4.50	4.50
Free space loss		(dB)	144.45	143.08	140.62	138.60	137.01	134.91	133.82
Atmospheric loss		(dB)	0.00	0.00	0.00	0.00	0.00	0.00	0.00
Polarisation loss		(dB)	0.45	0.45	0.45	0.45	0.45	0.45	0.45
Reflection and multipath	note 7	(dB)	1.00	1.00	1.00	1.00	1.00	1.00	1.00
Total propagation loss		(dB)	145.90	144.53	142.07	140.05	138.46	136.36	135.27
Ground station G/T (reference)	note 2	(dB/K)	-22.40	-22.40	-22.40	-22.40	-22.40	-22.40	-22.40
G/S axial ratio		(dB)	3.50	3.50	3.50	3.50	3.50	3.50	3.50
G/S pointing loss		(dB)	0.20	0.20	0.20	0.20	0.20	0.20	0.20
C/KT at receiver input		(dB/Hz)	68.10	69.47	71.93	73.91	73.94	73.92	73.94
Boltzmann's constant		(dBW/k Hz	-228.60	-228.60	-228.60	-228.60	-228.60	-228.60	-228.60
Bit rate		(dBHz)	48.57	48.57	48.57	48.57	48.57	48.57	48.57
Eb/No at receiver input		(dB)	19.53	20.90	23.36	25.34	25.37	25.35	25.37
Required Eb/No (BER=10 ⁻³)	note 13	(dB)	3.50	3.50	3.50	3.50	3.50	3.50	3.50
Coding loss due to frame sync	note 11	(dB)	0.47	0.47	0.47	0.47	0.47	0.47	0.47
Modulation degradation	note 8	(dB)	1.00	1.00	1.00	1.00	1.00	1.00	1.00
Receiver degradation	note 3	(dB)	2.00	2.00	2.00	2.00	2.00	2.00	2.00
System margin	note 4	(dB)	1.00	1.00	1.00	1.00	1.00	1.00	1.00
Available link margin		(dB)	11.56	12.93	15.39	17.37	17.40	17.38	17.41

note 1: minimum EIRP at S/C interface

note 2: reference G/T of the ground station; the actual figure mainly depends on the receiver, the level of man-made noise in the receiver location and the pointing of the antenna

note 3: allowed deviation from theoretical performance of the demodulator



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note 4: minimum system margin required by ESA

note 7: loss due to reflections from ground or from adjacent buildings

note 8: deviation from theoretical performance due to the modulation process, requirement by ESA

note 11: loss due the addition of synchronisation words in the interleaving protocol

note 13: theoretical Eb/No required to obtain a BER of 10⁻³ at the output of the Viterbi decoder. This BER requirement (due to the following R-S decoder) guarantees a virtually packet loss free link quality.



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Link ID	METOP LRPT - Omnidirectional antenna receiver								
Ground station	LRPT (0	G/T=-30.4)							
	1	T							
Frequency		(MHz)	137.90	137.90	137.90	137.90	137.90	137.90	
S/C altitude		(km)	850.00	850.00	850.00	850.00	850.00	850.00	
Slant range		(km)	2,256	1,858	1,473	1,227	963	850	
S/C view angle		(deg.ree)	59.28	56.00	49.82	42.52	26.18	0.00	
Data rate		(kbit/s)	72.00	72.00	72.00	72.00	72.00	72.00	
G/S view angle		(deg.ree)	13.00	20.00	30.00	40.00	60.00	90.00	
S/C EIRP	note 1	(dBW)	8.00	8.00	7.96	6.40	4.28	3.21	
S/C antenna axial ratio		(dB)	4.50	4.50	4.50	4.50	4.50	4.50	
Free space loss		(dB)	142.30	140.62	138.60	137.01	134.91	133.82	
Atmospheric loss		(dB)	0.00	0.00	0.00	0.00	0.00	0.00	
Polarisation loss		(dB)	0.47	0.47	0.47	0.47	0.47	0.47	
Reflection and multipath	note 7	(dB)	2.00	2.00	2.00	2.00	2.00	2.00	
Total propagation loss		(dB)	144.77	143.09	141.07	139.48	137.38	136.29	
Ground station G/T (reference)	note 2	(dB/K)	-30.40	-30.40	-30.40	-30.40	-30.40	-30.40	
G/S axial ratio		(dB)	3.50	3.50	3.50	3.50	3.50	3.50	
G/S pointing loss		(dB)	0.00	0.00	0.00	0.00	0.00	0.00	
C/KT at receiver input		(dB/Hz)	61.43	63.11	65.09	65.12	65.10	65.12	
Boltzmann's constant		(dBW/kHz)	-228.60	-228.60	-228.60	-228.60	-228.60	-228.60	
Bit rate		(dBHz)	48.57	48.57	48.57	48.57	48.57	48.57	
Eb/No at receiver input		(dB)	12.86	14.54	16.52	16.55	16.53	16.55	
Required Eb/No (BER=10 ⁻³)	note 13	(dB)	3.50	3.50	3.50	3.50	3.50	3.50	
Coding loss due to frame sync	note 11	(dB)	0.47	0.47	0.47	0.47	0.47	0.47	
Modulation degradation	note 8	(dB)	1.00	1.00	1.00	1.00	1.00	1.00	
Receiver degradation	note 3	(dB)	2.00	2.00	2.00	2.00	2.00	2.00	
System margin	note 4	(dB)	1.00	1.00	1.00	1.00	1.00	1.00	
Available link margin		(dB)	4.89	6.57	8.55	8.58	8.56	8.58	

note 1: minimum EIRP at S/C interface

note 2: reference G/T of the ground station; the actual figure mainly depends on the receiver and the level of man-made noise in the receiver location

note 3: allowed deviation from theoretical performance of the demodulator

note 4: minimum system margin required by ESA

note 7: loss due to reflections from ground or from adjacent buildings

note 8: deviation from theoretical performance due to the modulation process, requirement by ESA

note 11: loss due the addition of synchronisation words in the interleaving protocol



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note 13: theoretical Eb/No required to obtain a BER of 10⁻³ at the output of the Viterbi decoder. This BER requirement (due to the following R-S decoder) guarantees a virtually packet loss free link quality.

\		7.0							
Link ID	METOP HRPT								
Ground Station	HRPT (G/T=6.0)								
Frequency		(MHz)	1707.00	1707.00	1707.00	1707.00	1707.00	1707.00	1707.00
S/C altitude		(km)	850.00	850.00	850.00	850.00	850.00	850.00	850.00
Slant range		(km)	2889	2468	1858	1473	1227	963	850
S/C view angle		(degree.)	61.51	60.33	56.00	49.82	42.52	26.18	0.00
Data rate		(kbit/s)	3500.00	3500.00	3500.00	3500.00	3500.00	3500.00	3500.00
G/S view angle		(degree.)	5.00	10.00	20.00	30.00	40.00	60.00	90.00
S/C EIRP	note 1	(dBW)	9.10	9.10	8.23	6.21	4.65	2.55	1.46
S/C antenna axial ratio		(dB)	4.50	4.50	4.50	4.50	4.50	4.50	4.50
Free space loss		(dB)	166.30	164.93	162.47	160.45	158.86	156.76	155.67
Rain loss		(dB)	0.30	0.30	0.30	0.30	0.30	0.30	0.30
Polarisation Loss		(dB)	0.13	0.13	0.13	0.13	0.13	0.13	0.13
Reflection and multipath	note 7	(dB)	0.00	0.00	0.00	0.00	0.00	0.00	0.00
Total propagation loss		(dB)	166.73	165.36	162.90	160.88	159.29	157.19	156.10
Ground station G/T (reference)	note 2	(dB/K)	6.00	6.00	6.00	6.00	6.00	6.00	6.00
G/S axial ratio		(dB)	1.00	1.00	1.00	1.00	1.00	1.00	1.00
G/S pointing loss		(dB)	0.50	0.50	0.50	0.50	0.50	0.50	0.50
C/KT at receiver input		(dB/Hz)	76.47	77.84	79.43	79.43	79.46	79.46	79.46
Boltzmann's constant		(dBW/kHz)	-228.60	-228.60	-228.60	-228.60	-228.60	-228.60	-228.60
Bit rate		(dBHz)	65.44	65.44	65.44	65.44	65.44	65.44	65.44
Available Eb/No		(dB)	11.03	12.40	13.99	13.99	14.02	14.02	14.02
Required Eb/No (BER=10 ⁻³)	note 9	(dB)	4.00	4.00	4.00	4.00	4.00	4.00	4.00
Modulation degradation	note 3	(dB)	1.00	1.00	1.00	1.00	1.00	1.00	1.00
Receiver degradation	note 8	(dB)	2.00	2.00	2.00	2.00	2.00	2.00	2.00
System margin	note 4	(dB)	1.00	1.00	1.00	1.00	1.00	1.00	1.00
Available link margin		(dB)	3.03	4.40	5.99	5.99	6.02	6.02	6.02

note 1: minimum EIRP at S/C interface

note 2: reference G/T of the ground station

note 3: deviation from theoretical performance due to the modulation process; requirement by ESA

note 4: minimum system margin required by ESA

note 7: loss due to reflections from ground or from adjacent buildings

note 8: allowed deviation from theoretical performance of the demodulator



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note 9: theoretical Eb/No required to obtain a BER of 10⁻³ at the output of the Viterbi decoder. This BER requirement (due to the following R-S decoder) guarantees a virtually packet loss free link quality.

Annex 2. AVHRR/LRPT Data Compression system

A2.1 General

The AVHRR is a multi-spectral imager able to scan the earth in six spectral bands covered by five channels. The AVHRR scanning rate is 360 rpm producing one line per channel of earth view samples and calibration data every 1/6th of second.

The AVHRR samples are 10 bits wide and each earth view line contains 2048 pixels corresponding to a data rate of 2048(samples) x 5 (channels) x 6 (lines per second) x 10 (bits per sample), thus yielding an overall data rate of about 600 kbit/s.

The compression will operate on 10 bit input data. The output data will have the same resolution than the input data, i.e. the JPEG extended DCT-based process using 12 bit samples will be used. The AVHRR/LRPT compression system will deliver to the ground user 3 compressed channel out of the 5 channel provided by the AVHRR. The compression will operate on the 10 bit samples. The three compressed channels may be any combination of the channels available on board.

The data compression algorithm used for the AVHRR/LRPT is a modified version of the standard JPEG to adapt to a fixed compression ratio option and a continuous instrument operation mode (neither header nor trailer in the compressed stream)

The fixed compression ratio option is used to reduce the on board smoothing buffer size and to cope with a fixed average output data rate and to avoid risks of overflow.

The baseline of the JPEG algorithm is maintained, including the 8*8 sub-block extraction, quantization, zigzag reordering and Huffman coding [see for reference: Digital Compression and Coding of Continuous-tone Still Images, ISO/IEC CD10918-1, part 1, Draft, June 1991]



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A simple schematic of the algorithm is shown here below:

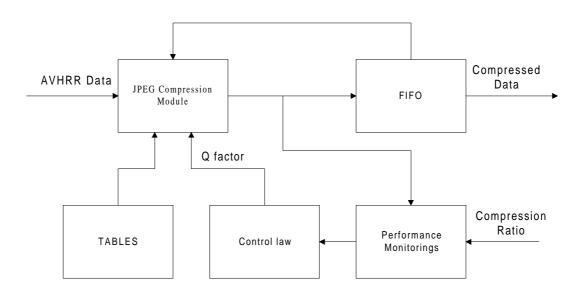


Figure A2.1.1 - Block diagram of the compression algorithm

The performance monitoring block computes of the actual compression ratio and compares it with the desired one. The result of the comparison is used to feed a nonlinear, empirical control-law that controls the Q factor.

The details of the compression algorithm are given in MO-TN-ESA-SY-0124 "LRPT Reference Compressor and Decompressor Technical Note".

In order to achieve a fixed compression ratio, the compressed stream is divided into segments containing a number of MCUs. For every segment the difference between the actual compression ratio and the desired one is computed (Delta CR) and the new quality factor is computed according to the control law, thus having a constant Q factor within the segment.

The Q factor is communicated to the decoder by means of one of the spare JPEG markers, along with the compressed data stream just between two MCUs. To allow the restart after an error (resynchronisation), the DC component of the first MCU of the first segment of every packet is not DPCM coded.



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A2.2 Data Format

A2.2.1 CCSDS Packets

In the on board Compression Unit (CU), the earth view samples and calibration data are treated in two different ways: The earth view samples of each selected channel *i* are grouped into stripes of 8 lines * 2048 pixels; each stripe is then compressed and formatted in a CCSDS packet having an APID associated to the channel *i*; calibration data instead are sub-sampled once every stripe and formatted in a non compressed format in a specific CCSDS packet.

There are seven different APIDs:

APID 64) coded data channel 1;

APID 65) coded data channel 2;

APID 66) coded data channel 3a;

APID 67) coded data channel 3b;

APID 68) coded data channel 4;

APID 69) coded data channel 5;

APID 70) calibration data.

The first three packets of the sequence will be any combination of the packets with APID 64, APID 65, either APID 66 or APID 67 (day or night), APID 68, APID 69. The sequence is closed by the packet with APID 70 and then is repeated continuously.



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A2.2.1.1 Compressed image packets (APID 64-69)

The following drawing illustrates the packet structure.

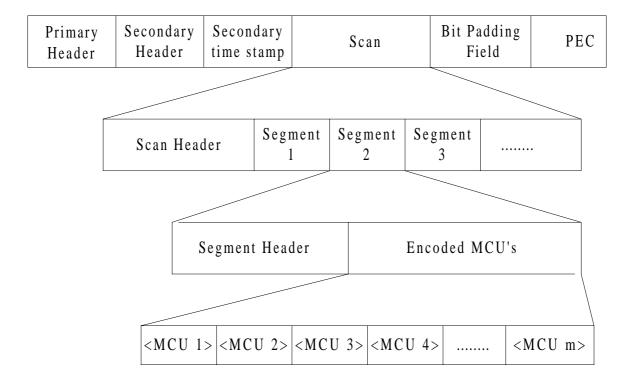


Figure A2.2.1.1.1 - Compressed packet structure

The Scan can contain any number (n) of Segments and each Segment can contain any number (m) of MCUs (Minimum Coded Units) provided that the global number of MCUs per packet is 2048 / 8 = 256 MCUs.

The decoder can recognize the beginning of a new Segment by means of its header and to decode the MCUs belonging to a Segment the quality factor specified in the Segment header must be used.

The tables can change from one CCSDS compressed packet to the other but not inside a single Scan. Quantization and Huffman tables to be used within the Scan are specified in the Scan header.

To allow the restart (re-synchronization) after an error, the DC component of the first MCU of the first Segment of every packet is not coded. This corresponds to setting the resynch interval to 256 (the number of MCUs per packet) in the JPEG standard.

The Bit Padding field is a variable length field and will allow the entire packet length to be an even number of bytes. Padding bits are all set to ones.

The secondary header will contain 8 bytes for the Time Tag. The secondary time stamp will



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contain 6 bytes for the additional time stamp.

Scan Header

At the beginning of each packet the Scan Header signals the decoder which tables are to be used in the decompression of the stripe. This field - see next figure - contains 3 fields of 4 bits each, thus allowing to use up to 16 different tables.

Qt (4 bits)	DC Ht (4 bits)	AC Ht (4 bits)
Qt (1 olts)	DC III (1 oits)	110 110 (1010)

Scan Header

Qt: index of the quantization table to be used in the scan;

DC Ht: index of the DC Huffman table to be used in the scan;

AC Ht: index of the AC Huffman table to be used in the scan.

The Qt field corresponds to the Tq field in the frame header as described in JPEG standard. Analogously, DC Ht and AC Ht correspond to the Td and Ta fields in the scan header.

Segment Header

The Segment Header can be found anywhere in the Scan stream, between two successive MCUs, and it is used to communicate to the decoder that a new quality factor is used in the following MCUs, until a new Segment Header is found.

As shown in the following drawing, the Segment Header is a JPEG marker used to allow the detection of this field by the decoder.

Segment header

QFM: quality factor mark. Is one of the reserved JPEG markers, in this application it is set to 0xFFF0;

O factor: it is the quality factor (or quantization factor) to be used within the Segment.



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A2.2.1.2 Calibration Packet (APID 70)

The following drawing illustrates the calibration packet structure.

Primary Header	Secondary Secondar Header time stan		•	Calibratio	on Data	Reserved for future use (34 bits)		PEC
Channel 1 ancillary data (230 bits)	Channel 2 ancillary data (230 bits)		or cha ancill	nnel 3a annel 3b ary data O bits)	ancilla	nnel 1 ary data bits)	Channel 1 ancillary data (230 bits)	

Figure A2.2.1.2.1 - Calibration packet structure

The secondary header contains a Time Stamp field of 6 octets.

The calibration data field is composed by a field reserved for future use and five fields of 230 bits for the ancillary data of each channel.



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A2.3 System Architecture

The ground system is mainly composed of two subsystems: the Signal Conditioning subsystem and the Digital Processing subsystem.

The Signal Conditioning subsystem, knowing the frame synchronisation pattern and the frame length can detect the CADUs and then can de-randomise the CVCDUs. The CVCDUs are then de-interleaved and R-S decoded to remove channel errors. The valid VCDUs are sent to the packet demultiplexer that sorts out the AVHRR packets that are stored in the host computer's hard disk either in DMA or under program control.

Administrative packets are first decoded to enable a correct setting of the decompression algorithm.

During the mission lifetime, it is possible that predefined quantization and Huffman tables (available to both space and ground systems) are updated or expanded. These new tables are broadcasted to the users by means of the Administrative messages.

Once the packets have been stored on the disk, the decompression process can take place off-line for visualisation purposes.



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A2.4 Software decoding

CCSDS Packets will be first grouped and archived. After the packets have been properly conditioned, visualisation can be performed by commercial or public domain tools.

The CCSDS packets conditioning process consists mainly in the decompression of several Encoded Channel Stripes belonging to packets having the same AP ID so that the raw data of a specific channel can be reconstructed. Then a conversion to a known image format can take place.

More in detail the conditioning consists of:

- stripping-off the header and trailer information of CCSDS packets related to a specific channel and observation time. For instance, to create a 2048*2048 pixel image of the AVHRR channel 1, a set of 256 consecutive packets with APID 64 must be considered
 - reading the Administrative Messages to seek for new Quantization or Huffman tables
- for every new packet the Scan Header is read out to select the proper tables to be used within the Scan. If the specified tables are different from the ones predefined they must have been defined in the Administrative Messages
- detecting the Segment Headers so that a new Segment can be identified. The Segment Header contains the Q factor value which must be used to re-scale the quantization table addressed by the Qt field in the Scan Header
 - decompressing the MCUs according to JPEG compression standard

The use of a lossy compressed format (such as JPEG) for the reconstructed images is not advisable because even with the highest quality set it would lead to some added loss beside the one due to the on-board compression algorithm.





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Annex 3. LIST OF ACRONYMS

AM Amplitude Modulation

AMSU Advanced Microwave Sounding Unit

APid Application Process identifier

ASCAT Advanced Scatterometer

AVHRRAdvanced Very High Resolution Radiometer

BER Bit Error Rate

CADU Channel Access Data Unit

CCSDS Consultative Committee for Space Data Systems

CRC Cyclic Redundancy Checksum CVCDUCoded Virtual Channel Data Unit

DCS Data Collection System

 E_b/N_0 Bit energy / noise density

EIRP Equivalent Isotropic Radiated Power

FEC Forward Error Correction

GRAS GNSS Receiver for Atmospheric Sounding

G/S Ground Station

G/T Figure of merit (antenna gain ./. system noise temperature)

HIRS High resolution InfraRed Sounder HRPT High Resolution Picture Transmission

IASI Infrared Atmospheric Sounding Interferometer

IN_PDUInsert service_ Protocol Data Unit

IN_SDUInsert _Service Data Unit

LRPT Low Resolution Picture Transmission

MHS Microwave Humidity Sounder M_PDU Multiplex_ Protocol Data Unit M_SDU Multiplex _Service Data Unit

NRZ-L Nonreturn-to-zero Type L

GOME Global Ozone Monitoring Instrument

PCA_PDU Physical Channel Access_Protocol Data Unit







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PN Pseudo Noise

QPSK Quadrature Phase Shift Keying

RF Radio Frequency RHC Right Hand Circular

SEM Space Environment Monitor

S/C Spacecraft

TBC To Be Confirmed

TBUS TIROS Bulletin United States

TBD To Be Defined

UTC Coordinated universal time

VC Virtual Channel

VCA Virtual Channel Access

VCA_SDU Virtual Channel Access _Service Data Unit

VCLC Virtual Channel Link Control

VCDU Virtual Channel Data Unit

VCDU-ID Virtual Channel Data Unit Identifier